



**NTNU – Trondheim**  
Norwegian University of  
Science and Technology

## Examination paper for TFY4280 Signal Processing

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**Examination date:** 15.05.2015

**Examination time (from-to):** 0900 - 1300

**Permitted examination support material:**

- Simple calculator (according to NTNU exam regulations)
- K. Rottmann: Matematisk formelsamling
- Barnett and Cronin: Mathematical formulae
- Carl Angell og Bjørn Ebbe Lian: Fysiske størrelser og enheter, navn og symboler

**Other information about the exam paper:**

- Language: English
- Number of pages (including this page and attachments): 12
- Answer must be written in English or Norwegian. Number of points given to each sub-question is given in bold font. The maximum score for the exam is **100p**.
- **Attachment:** 2 pages with transform tables and properties.

**Checked by:**

**Date:**

**Signature:**

**Q1 (30p)**

- A) (10p)** Describe the concept of *transfer function* for continuous-time and discrete-time LTI systems. What properties of the LTI system allows you to use transfer function to determine output signal for a given input signal. Explain.
- B) (10p)** Find the transfer functions and the corresponding system equations for continuous-time and discrete-time LTI systems described by the impulse response functions given below

$$\begin{aligned} h_1(t) &= \varepsilon(t)e^{-at} \sin(\omega_0 t) \\ h_2[n] &= \alpha\delta[n] + (1 - \alpha)h[n - 1] \\ h_3[n] &= [\underline{10} \ 9 \ 8 \ 7] \end{aligned}$$

- C) (10p)** How can one use transfer function to describe the frequency response of a LTI system? How are these two concepts connected? Briefly explain how to calculate frequency responses for systems with impulse given above (you do not have to do the full calculation). Where necessary, use sampling time  $t_s = \frac{2\pi}{100}$ s.

**Q2 (30p)**

- A) (10p)** Find the unilateral ( $n \geq 0$ ) z-transform of

$$x[n] = 5 \cos[3n]$$

- B) (10p)** Determine the convolution

$$y(t) = e^{-at} \varepsilon(t) * \varepsilon(t) \tag{1}$$

using the Fourier transform method.

- C) (10p)** A system with a transfer function

$$H(s) = \frac{s - 1}{s^2 + 3s + 2} \tag{2}$$

is excited by white noise with power density  $N_0$  giving an output signal  $y(t)$ . Determine the ACF  $\varphi_{yy}(\tau)$ , the mean  $\mu_x$  and the variance  $\sigma_x^2$  of the output signal  $y(t)$

**Q3 (20p)**

A) (10p) Show that:

$$\mathcal{L}\{t \cdot x(t)\} = -\frac{dX(s)}{ds}$$

B) (10p) Use above property to calculate output of LTI system where input  $x(t) = te^{-9t}$  defined for  $t > 0$  and the impulse response is given by:

$$H(s) = \frac{1}{(s + 10)}$$

**Q4 (20p)**

A) (10p) Explain the concept of **discrete frequency** by considering Fourier transform of a sampled signal:

$$x_s(t) = \sum_{n=-\infty}^{\infty} x(nT_s)\delta(t - nT_s)$$

B) (10p) Describe the difference between DTFT and DFT. Show how to calculate both transforms for a signal defined by

$$x[n] = \begin{cases} 1 & 0 \leq n < 10 \\ 0 & \text{otherwise} \end{cases}$$

NOTE: Here it is sufficient to express the transforms in terms of a power series.

**Appendix B.1 Bilateral Laplace Transform Pairs**

$x(t)$	$X(s) = \mathcal{L}\{x(t)\}$	ROC
$\delta(t)$	1	$s \in \mathbf{C}$
$\varepsilon(t)$	$\frac{1}{s}$	$\text{Re}\{s\} > 0$
$e^{-at}\varepsilon(t)$	$\frac{1}{s+a}$	$\text{Re}\{s\} > \text{Re}\{-a\}$
$-e^{-at}\varepsilon(-t)$	$\frac{1}{s+a}$	$\text{Re}\{s\} < \text{Re}\{-a\}$
$t\varepsilon(t)$	$\frac{1}{s^2}$	$\text{Re}\{s\} > 0$
$t^n\varepsilon(t)$	$\frac{n!}{s^{n+1}}$	$\text{Re}\{s\} > 0$
$te^{-at}\varepsilon(t)$	$\frac{1}{(s+a)^2}$	$\text{Re}\{s\} > \text{Re}\{-a\}$
$t^n e^{-at}\varepsilon(t)$	$\frac{n!}{(s+a)^{n+1}}$	$\text{Re}\{s\} > \text{Re}\{-a\}$
$\sin(\omega_0 t)\varepsilon(t)$	$\frac{\omega_0}{s^2 + \omega_0^2}$	$\text{Re}\{s\} > 0$
$\cos(\omega_0 t)\varepsilon(t)$	$\frac{s}{s^2 + \omega_0^2}$	$\text{Re}\{s\} > 0$
$e^{-at}\cos(\omega_0 t)\varepsilon(t)$	$\frac{s+a}{(s+a)^2 + \omega_0^2}$	$\text{Re}\{s\} > \text{Re}\{-a\}$
$e^{-at}\sin(\omega_0 t)\varepsilon(t)$	$\frac{\omega_0}{(s+a)^2 + \omega_0^2}$	$\text{Re}\{s\} > \text{Re}\{-a\}$
$t\cos(\omega_0 t)\varepsilon(t)$	$\frac{s^2 - \omega_0^2}{(s^2 + \omega_0^2)^2}$	$\text{Re}\{s\} > 0$
$t\sin(\omega_0 t)\varepsilon(t)$	$\frac{2\omega_0 s}{(s^2 + \omega_0^2)^2}$	$\text{Re}\{s\} > 0$

**Appendix B.3 Fourier Transform Pairs**

$x(t)$	$X(j\omega) = \mathcal{F}\{x(t)\}$
$\delta(t)$	1
1	$2\pi\delta(\omega)$
$\delta'(t)$	$j\omega$
$\frac{1}{T} \text{III} \left( \frac{t}{T} \right)$	$\text{III} \left( \frac{\omega T}{2\pi} \right)$
$\varepsilon(t)$	$\pi\delta(\omega) + \frac{1}{j\omega}$
$\text{rect}(at)$	$\frac{1}{ a } \text{si} \left( \frac{\omega}{2a} \right)$
$\text{si}(at)$	$\frac{\pi}{ a } \text{rect} \left( \frac{\omega}{2a} \right)$
$\frac{1}{t}$	$-j\pi \text{sign}(\omega)$
$\text{sign}(t)$	$\frac{2}{j\omega}$
$e^{j\omega_0 t}$	$2\pi\delta(\omega - \omega_0)$
$\cos(\omega_0 t)$	$\pi[\delta(\omega + \omega_0) + \delta(\omega - \omega_0)]$
$\sin(\omega_0 t)$	$j\pi[\delta(\omega + \omega_0) - \delta(\omega - \omega_0)]$
$e^{-\alpha t }, \alpha > 0$	$\frac{2\alpha}{\alpha^2 + \omega^2}$
$e^{-a^2 t^2}$	$\frac{\sqrt{\pi}}{a} e^{-\frac{\omega^2}{4a^2}}$

**Appendix B.2 Properties of the Bilateral Laplace Transform**

$x(t)$	$X(s) = \mathcal{L}\{x(t)\}$	ROC
Linearity $Ax_1(t) + Bx_2(t)$	$AX_1(s) + BX_2(s)$	ROC $\supseteq$ $\text{ROC}\{X_1\} \cap \text{ROC}\{X_2\}$
Delay $x(t - \tau)$	$e^{-s\tau}X(s)$	not affected
Modulation $e^{at}x(t)$	$X(s - a)$	$\text{Re}\{a\}$ shifted by $\text{Re}\{a\}$ to the right
'Multiplication by $t$ ', Differentiation in the frequency domain $tx(t)$	$-\frac{d}{ds}X(s)$	not affected
Differentiation in the time domain $\frac{d}{dt}x(t)$	$sX(s)$	ROC $\supseteq$ $\text{ROC}\{X\}$
Integration $\int_{-\infty}^t x(\tau)d\tau$	$\frac{1}{s}X(s)$	$\text{ROC} \supseteq \text{ROC}\{X\}$ $\cap \{s : \text{Re}\{s\} > 0\}$
Scaling $x(at)$	$\frac{1}{ a }X\left(\frac{s}{a}\right)$	ROC scaled by a factor of $a$

**Appendix B.4 Properties of the Fourier Transform**

	$x(t)$	$X(j\omega) = \mathcal{F}\{x(t)\}$
Linearity	$Ax_1(t) + Bx_2(t)$	$AX_1(j\omega) + BX_2(j\omega)$
Delay	$x(t - \tau)$	$e^{-j\omega\tau}X(j\omega)$
Modulation	$e^{j\omega_0 t}x(t)$	$X(j(\omega - \omega_0))$
'Multiplication by $t$ ' Differentiation in the frequency domain $tx(t)$	$tx(t)$	$-\frac{dX(j\omega)}{d(j\omega)}$
Differentiation in the time domain $\frac{dx(t)}{dt}$	$\frac{dx(t)}{dt}$	$j\omega X(j\omega)$
Integration	$\int_{-\infty}^t x(\tau)d\tau$	$X(j\omega) \left[ \pi\delta(\omega) + \frac{1}{j\omega} \right]$ $= \frac{1}{j\omega}X(j\omega) + \pi X(0)\delta(\omega)$
Scaling	$x(at)$	$\frac{1}{ a }X\left(\frac{j\omega}{a}\right), a \in \mathbb{R} \setminus \{0\}$
Convolution	$x_1(t) * x_2(t)$	$X_1(j\omega) \cdot X_2(j\omega)$
Multiplication	$x_1(t) \cdot x_2(t)$	$\frac{1}{2\pi}X_1(j\omega) * X_2(j\omega)$
Duality	$x_1(t)$ $x_2(jt)$	$x_2(j\omega)$ $2\pi x_1(-\omega)$
Symmetry relations	$x(-t)$ $x^*(t)$ $x^*(-t)$	$X(-j\omega)$ $X^*(-j\omega)$ $X^*(j\omega)$
Parseval theorem	$\int_{-\infty}^{\infty}  x(t) ^2 dt$	$\frac{1}{2\pi} \int_{-\infty}^{\infty}  X(j\omega) ^2 d\omega$

**Appendix B.6 Properties of the  $z$ -Transform**

Property	$x[k]$	$X(z)$	ROC
Linearity	$ax_1[k] + bx_2[k]$	$aX_1(z) + bX_2(z)$	$\text{ROC} \supseteq \text{ROC}\{X_1\} \cap \text{ROC}\{X_2\}$
Delay	$x[k - \kappa]$	$z^{-\kappa}X(z)$	$\text{ROC}\{x\}$ ; separate consideration of $z = 0$ and $z \rightarrow \infty$
Modulation	$a^k x[k]$	$X\left(\frac{z}{a}\right)$	$\text{ROC} = \left\{ z \mid \frac{z}{a} \in \text{ROC}\{x\} \right\}$
Multiplication by $k$	$kx[k]$	$-z \frac{dX(z)}{dz}$	$\text{ROC}\{x\}$ ; separate consideration of $z = 0$
Time inversion	$x[-k]$	$X(z^{-1})$	$\text{ROC} = \{z \mid z^{-1} \in \text{ROC}\{x\}\}$
Convolution	$x_1[k] * x_2[k]$	$X_1(z) \cdot X_2(z)$	$\text{ROC} \supseteq \text{ROC}\{x_1\} \cap \text{ROC}\{x_2\}$
Multiplication	$x_1[k] \cdot x_2[k]$	$\frac{1}{2\pi j} \oint X_1(\zeta) X_2\left(\frac{z}{\zeta}\right) \frac{1}{\zeta} d\zeta$	multiply the limits of the ROC

**Appendix B.5 Two-sided  $z$ -Transform Pairs**

$x[k]$	$X(z) = \mathcal{Z}\{x[k]\}$	ROC
$\delta[k]$	1	$z \in \mathbf{C}$
$\varepsilon[k]$	$\frac{z}{z-1}$	$ z  > 1$
$a^k \varepsilon[k]$	$\frac{z}{z-a}$	$ z  >  a $
$-a^k \varepsilon[-k-1]$	$\frac{z}{z-a}$	$ z  <  a $
$k\varepsilon[k]$	$\frac{z}{(z-1)^2}$	$ z  > 1$
$ka^k \varepsilon[k]$	$\frac{az}{(z-a)^2}$	$ z  >  a $
$\sin(\Omega_0 k) \varepsilon[k]$	$\frac{z \sin \Omega_0}{z^2 - 2z \cos \Omega_0 + 1}$	$ z  > 1$
$\cos(\Omega_0 k) \varepsilon[k]$	$\frac{z(z - \cos \Omega_0)}{z^2 - 2z \cos \Omega_0 + 1}$	$ z  > 1$